

ENERGY AUDIT OF INDUCTION MOTORS

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ABSTRACT

Conducting energy audits of asynchronous motors is a key factor in achieving energy savings. This paper describes the calculation methods.

Keywords: frequency-controlled asynchronous motor, electric drive, torque.

The most common case is the frequency-controlled asynchronous motorized automated electric drive, while other methods performed at a certain frequency are the special case h of frequency control.

We describe the analysis and calculation method of asynchronous motor operation with minimum power dissipation in frequency-adjusted automated electric drive systems.

To calculate the operating and adjustment characteristics of frequencyadjustable electric drives and to analyze the adjustment characteristics, we present the analytical relations expressed by the magnetic flux and determine that the magnetic power dissipation in electric motors is related to determine the optimal value of the magnetic flux. To simplify the analytical relationship obtained for the T-shaped equivalent circuit and vector diagram, we give only for the harmonics k =1. The relative value of the magnetic flux of an induction motor

$$\varphi = \frac{\Phi}{\Phi}$$
,

while the relative values of frequency and torque

$$F = \frac{f}{f_{\kappa}}, \mu = \frac{M}{M_{\mu}}$$

and torque.

The applied current of the rotor:

$$I_{PF\varphi} = \sqrt{\frac{P_{\Im M.H}}{m_1 r_P} \, \beta \varphi}$$

 $I_{PF\varphi}=\sqrt{\frac{P_{\Im M.H}}{m_1r_P}}\,\beta\varphi,$ this here R em.n - nominal electromagnet q uvvat , m 1 - of the stator phases soni;



$$\beta\varphi = d\varphi^2 - \sqrt{(a\varphi^2)^2 - c}$$

$$a = \frac{m_1 E_{cH} r^1 p}{2 P_{\Im M.H} x_p^{12}} \varphi = \frac{r^{12} p}{x_p^{12}},$$
 absolute slip;

E sn - nominal value of stator EMF.

Magnetizer current

$$I_{OF,\varphi} = \frac{E_{OH}F\varphi}{\sqrt{r_{O1}^2 + x_{O1}^2 \gamma}}.$$

Magnetizer contour active and inductive Resistance (3.2) equation dan:

$$r_{OF,\varphi} = \frac{r_{\mu}F - \sqrt{(r_{\eta}F - 4x_{OF,\varphi})^{2}}}{2}$$

$$x_{OF,\varphi} = F\sqrt{\frac{E_{C.H.}^{2}\varphi^{2}}{I_{O\varphi}^{2}} - \left(\frac{\Delta P_{cm.H}\varphi^{2}}{m_{1}I_{OF}^{2}}\right)}$$

henceforth

 $^{\Delta}R$ sm.n. - nominal losses in stall I OP - F = When 1 b (magnetization) curves from the line $K=1,\!315$ -

$$I_{CF,\varphi} = E_{C.H.\varphi} \sqrt{\frac{(x_{OF,\varphi} + x_P^1 F)^2 + (r_{OF,\varphi} + \frac{r_P^1 F}{\beta \gamma})^2}{(r_{OF,\varphi} + x_{OF,\varphi}^2)(\frac{r_P^{12}}{\beta^2 \varphi} + x_P^{12})}},$$

Stator current

Slip
$$s_{F,\varphi} = \frac{\beta \varphi}{F}$$
.

Electromagnetic losses:

$$\Delta P_{\Im M.F.\varphi} = m_1 r_c E_{C.H.\varphi}^2 \frac{(x_{OF.\varphi} + x_{\mu F}^1)^2 + \left(r_{OF.\varphi} + \frac{r_P F}{\beta \varphi}\right)^2}{(r_{OF.\varphi}^2 + x_{OF.\varphi}^2) \left(\frac{r_P^{12}}{\beta^2 \varphi} + x_P^{12}\right)} + \Delta P_{\Im M.H} \beta_{\gamma} + \Delta P_{cm.H} \varphi^2 F^K.$$

General waste:



$$\sum \Delta P_{F,\varphi} = E_{C.H.\varphi}^2 \left(m_1 r_c + \frac{\Delta P_{\kappa y u...H}}{I_{C.H}^2} \right) \frac{(x_{OF.\varphi} + x_P^1 F)^2 + \left(r_{OF.\varphi} + \frac{r_P^1 F}{\beta \varphi} \right)^2}{(r_{OF.\varphi}^2 + x_{OF.\varphi}^2) \left(\frac{r^1 P}{\beta^2 \varphi} + x_P^{12} \right)} + \Delta P_{\Im M.H} \beta_{\gamma} + \frac{r_{\rho}^2 F}{\beta^2 \varphi} + \frac{r_{\rho}^2 F}{\beta^2 \varphi} \right) \frac{(x_{OF.\varphi} + x_P^2 F)^2 + \left(r_{OF.\varphi} + \frac{r_{\rho}^2 F}{\beta \varphi} \right)^2}{(r_{OF.\varphi}^2 + x_{OF.\varphi}^2) \left(\frac{r^1 P}{\beta^2 \varphi} + x_P^{12} \right)} + \Delta P_{\Im M.H} \beta_{\gamma} + \frac{r_{\rho}^2 F}{\beta^2 \varphi} + \frac{r_{\rho}^2 F}{\beta^2 \varphi} \right) \frac{(x_{OF.\varphi} + x_P^2 F)^2 + \left(r_{OF.\varphi} + \frac{r_{\rho}^2 F}{\beta \varphi} \right)^2}{(r_{OF.\varphi} + x_{OF.\varphi}^2) \left(\frac{r^1 P}{\beta^2 \varphi} + x_P^{12} \right)} + \Delta P_{\Im M.H} \beta_{\gamma} + \frac{r_{\rho}^2 F}{\beta^2 \varphi} + \frac{r_{\rho}^2 F}{\beta^2 \varphi} \right) \frac{(x_{OF.\varphi} + x_P^2 F)^2 + \left(r_{OF.\varphi} + \frac{r_{\rho}^2 F}{\beta \varphi} \right)^2}{(r_{OF.\varphi} + x_{OF.\varphi}^2) \left(\frac{r^1 P}{\beta^2 \varphi} + x_P^{12} \right)} + \Delta P_{\Im M.H} \beta_{\gamma} + \frac{r_{\rho}^2 F}{\beta^2 \varphi} + \frac{r_{\rho}^2 F}{\beta^2 \varphi} \right) \frac{(x_{OF.\varphi} + x_P^2 F)^2 + r_{\rho}^2 F}{(r_{OF.\varphi} + x_P^2 F)^2 + r_{\rho}^2 F}$$

$$\Delta P_{cm,H} \varphi^2 F^K + M_H \omega_H (F - \beta_{\gamma}),$$

where the I_{CH} , ω_H , M_H , $\Delta P_{syu.u}$ -nominal values of stator current, synchronous speed, mechanical torque, and additional losses.

$$P_{\partial F.\varphi} = M_H \omega_H (F - \beta_{\gamma}),$$

this here M n - of the engine nominal torque on the shaft . Required power

$$P_{nF..\varphi} = E_{OH}^{2} \left(m_{1} r_{c} + \frac{\Delta P_{\kappa y u..u}}{I_{CM}^{2}} \right) \varphi^{2} \frac{\left(x_{OF.\varphi} + x_{\mu F}^{1} \right)^{2} + \left(r_{OF.\varphi} + \frac{r_{p}^{12} F}{\beta \varphi} \right)}{\left(r_{OF.\varphi}^{2} + x_{OF.\varphi}^{2} \right) \left(\frac{r_{p}^{12}}{\beta^{2} \varphi} + x_{p}^{12} \right)} + \Delta P_{\Im M.H} F + \Delta P_{cm.u} \varphi^{2} F^{K}$$

Expression of CE and power factor of motor indicators:

$$\eta_{F,\phi} = \frac{P_{\partial F\phi}}{P_{\eta F,\phi}} = \frac{M_{\partial H} \omega_{oH} (F - \beta_{\phi}) (x_{OF,\phi} + x_P^1 F)^2 + (r_{OF,\phi} + \frac{r_P^1 F}{\beta_{\phi}})^2}{\Delta P_{\partial M.H} F + \Delta P_{cm.H} \varphi^2 F^2 + E_{CH}^2 \varphi^2 (m_1 r_C + \frac{\Delta P_{\partial OB.H}}{I_{CH}^2}) (r_{OF,\phi}^2 + x_{OF,\phi}^2) (\frac{2^1_{PF}}{\beta \varphi} + x_P^{12})}$$

$$\cos \varphi_{F,\varphi} = \frac{P_{nF,\varphi}}{m_1 U I_{CF,\varphi}} = \left[\frac{E_{CH\varphi}(m_1 r_c + \frac{\Delta P_{\Im O E,H}}{I_{CH}^2})(x_{OF,\varphi} + x_{p\varphi}^1)^2 + (r_{OF,\varphi} + \frac{r_{pF}^1}{\beta \varphi})^2}{m_1 U(r_{OF,\varphi} + x_{OF,\varphi}^2)(\frac{r_p^{12}}{\beta \varphi} + x_p^{12})} + \frac{\Delta P_{\Im M,H} F + \Delta P_{cm,H} \varphi^2 F^K}{m_1 U E_{CH,\varphi}} \right] x$$

$$x \sqrt{\frac{(r_{OF,\varphi}^2 + x_{OF,\varphi}^2) \left(\frac{r_p^{12}}{\beta \gamma} + x_\rho^{12}\right)}{(x_{OF,\varphi} + x_p^1 F)^2 + \left(r_{OF,\varphi} + \frac{r_p^{12} F}{\beta \varphi}\right)^2}}$$

Energy indicator

$$\eta_{F,\varphi} \cos \varphi_{F,\varphi} = \frac{P_{\partial F,\varphi}}{m_1 U I_{CF,\varphi}} = \frac{M_{\partial H} \omega_{OH} (F - \beta \varphi)}{m_1 U E_{CH} \varphi} x \sqrt{\frac{(r_{OF,\varphi}^2 + x_{OF,\varphi}^2) \left(\frac{r_p^{12}}{\beta_{\varphi}^2} + x_p^{12}\right)}{(x_{OF,\varphi} + x_{PF}^1) + \left(r_{OF,\varphi} + \frac{r_p^1 F}{\beta \varphi}\right)^2}}$$

The voltage U corresponding to the defined values of F and s φ



can be defined as follows:

$$U = \sqrt{2x_C^2 F^2 I_{CF,\phi}^2 - A_{F,\phi} + (2x_C^2 F I_{CF,\phi}^2 - A_{F,\phi})^2 - A_{F,\phi}^2 - \frac{4}{m_1^2} x_C^2 F^2 P_{PF,\phi}^2}$$

here
$$A_{F,\varphi} = I_{CF,\varphi}^2 (x_C^2 F^2 + r_C^2) - E_{CH}^2 F^2 \varphi^2 - \frac{2}{m} r_C P_{nF,\varphi}$$
.

The optimum value of current for different frequencies F is sufficient for g opt level of accuracy (error not greater than 2%) analytical without calculations method, DR can be determined without studying the EMF, X = SH(X) function.

In this case, we assume that the square of the stator current of an induction motor is equal to the sum of the squares of the applied current of the rotor and the magnetizing current.

$$I_{CF,\varphi}^2 = I_{P\varphi}^{12} + I_{O\varphi}^2$$

Of the rotor listed toki esa okimga reverse is proportional to:

$$I_{P\varphi}^{1} = \frac{\Delta P_{\Im M.H}}{m_{1} E_{CH} \varphi},$$
(14)

To express the square of the magnetizing current by current, we use the formula:

$$I_{O\varphi}^2 = I_{OH}^2 \frac{\gamma^2}{K_M - (K_M - 1)^2 \varphi},$$

where K M is the selection coefficient of the curve I 2 OX to be more accurate.

Based on the above initial cases, we obtain an approximate expression of electromagnetic losses:

here
$$B = (r_c + r_p^1)\Delta P_{\Im M.H} / m_1 E_{CH}^2 : C = 3r_C^2 I_{OH}^2 : D = \Delta P_{CTH}$$
.

by taking some addition from the expression and making it equal to zero, making some changes:

$$\varphi^2 + \varepsilon \varphi^2 + c_{\varphi} \varphi^2 + d_F \varphi^2 + e_{\varphi} = 0$$
, (17)

here

$$e = \frac{2K}{1 - K_{\mu}}; c_F = \frac{cr_{\mu} + DF^k K_{\mu}^2 - B(K_{\mu} - 1)}{DF^K (K_{\mu} - 1)^2};$$

$$d_F = \frac{2BK_{\mu}}{DF^k(K_{\mu}-1)}; e_F = \frac{B}{DF^k} \left(\frac{K_{\mu}^2}{K_{\mu}-1}\right);$$



Solving Equation (3.17), we obtain the general analytical expression of the optimal current, in which the power dissipated in an asynchronous motor in frequency-controlled systems is the smallest, and the CE is the highest:

$$\begin{split} \varphi_{\rm OHT} &= \sqrt{\frac{\varepsilon + A}{4}} + \sqrt{\left(\frac{\varepsilon + A}{4}\right)^2 - V \frac{\varepsilon \varphi - dF}{A}}, \\ A &= \sqrt{8\varphi + \varepsilon^2 - 4c_F}; \varphi = \sqrt[3]{-q + \sqrt{q^2 - p^2}} + \sqrt[3]{-q - \sqrt{q^2 + p^3}} + \frac{c_F}{6}, \\ q &= -\left(\frac{C_F}{6}\right)^3 + \frac{C_F (4d_F - \varepsilon^2) - d_F^2}{16}, p = -\left(\frac{C_F}{6}\right)^2; \end{split}$$
 here

by substituting the values obtained into, it is possible to obtain in the optimal mode the values of the quantities and parameters of interest to us, with the electromagnetic loss being the smallest (minimum).

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